



Article

Noncoherent Decision Fusion over Fading Hybrid MACs in Wireless Sensor Networks

Shoujun Liu¹, Kehao Wang ¹, Kezhong Liu² and Wei Chen ^{1,*}

- School of Information Engineering, Wuhan University of Technology, Wuhan 430070, China; lsj@whut.edu.cn (S.L.); kehao.wang@whut.edu.cn (K.W.)
- School of Navigation, Wuhan University of Technology, Wuhan 430063, China; kzliu@whut.edu.cn
- * Correspondence: greatchen@whut.edu.cn; Tel.: +86-137-0710-3937

Received: 11 December 2018; Accepted: 26 December 2018; Published: 1 January 2019



Abstract: In this paper, we consider the problem of decision fusion for noncoherent detection in a wireless sensor network. Novel to the current work is the integration of the hybrid multi-access channel (MAC) in the fusion rule design. We assume that sensors transmit their local binary decisions over a hybrid MAC which is a composite of conventional orthogonal and nonorthogonal MACs. Under Rayleigh fading scenario, we present a likelihood ratio (LR)-based fusion rule, which has been shown to be optimal through theoretical analysis and simulation. However, it requires a large amount of computation, which is not easily implemented in resource-constrained sensor networks. Therefore, three sub-optimal alternatives with low-complexity are proposed, namely the weighed energy detector (WED), the deflection-coefficient-maximization (DCM), and the two-step (TS) rules. We show that when the channel signal-to-noise ratio (SNR) is low, the LR-based fusion rule reduces to the WED rule; at high-channel SNR, it is equivalent to the TS rule; and at moderate-channel SNR, it can be approached closely by the DCM rule. Compared with the conventional orthogonal and nonorthogonal MACs, numerical results show that the hybrid MAC with the proposed fusion rules can improve the detection performance when the channel SNR is medium.

Keywords: decision fusion; distributed detection; hybrid MAC; wireless sensor networks

1. Introduction

Wireless sensor networks (WSNs) consist of a large number of geographically distributed sensors that have limited resources, such as energy, processing capabilities, and communication bandwidth [1]. With sensor collaboration, potentially powerful networks can be constructed to accomplish certain tasks, such as environment monitoring, surveillance, health care, and home automation [2–9].

A prevailing model used for such applications is the orthogonal multi-access channel (MAC) model wherein sensors make local decisions based on their observations and subsequently transmit these decisions in parallel to a fusion center (FC). Numerous researchers have focused on the problem of distributed detection (decision fusion) over orthogonal MACs [10–21].

Another type of MACs from sensors to the FC is the nonorthogonal MAC [22]. In this case, multiple sensors are allowed to communicate with an FC through the same channel. Using this scheme, bandwidth requirement or detection delay can be significantly reduced. However, in order to realize the nonorthogonal MAC, channel information needs to be fed back from the FC to each sensor so that synchronization among all the sensors can be established. Under the assumption of perfect synchronization, the type-based multiple-access scheme is studied in [23–25]. In [26], it is shown that the nonorthogonal MAC can achieve the same asymptotic performance as centralized detection. When wireless channels are subject to fading, in [27,28], the optimal likelihood ratio (LR) fusion rule is shown to be equivalent to a simple energy test under both Rayleigh and Rician fading channels.

Sensors **2019**, 19, 120 2 of 15

Further, decision fusion for WSNs with space diversity has been studied in [29–33]. Obviously, space diversity can effectively suppress the channel fading and hence lead to good performance. However, to employ space diversity, we need multiple antennas to be equipped at the FC. Due to the size constraint of sensors, this could be impractical for many WSN applications [34]. An alternative for space diversity is the cooperative transmission scheme [35,36], in which multiple relays are used to assist the sensors to send their decisions to the FC.

Recently, the idea of hybrid MACs with sensor grouping has emerged in [37]. For hybrid MACs, the orthogonal MAC is adopted across different groups, whereas the nonorthogonal MAC is used for sensors within each group. A flexible trade-off between these two MAC schemes can therefore be obtained by changing the number of groups and the number of sensors in each group. Using coherent detection and assuming channel gains are available, the hybrid MAC is shown to provide more performance choices than orthogonal and nonorthogonal MACs [38]. The use of coherent detection requires the knowledge of channel phrase. However, acquiring this phrase information usually results in additional training overhead, which may be too costly for resource-constrained sensor networks. Moreover, due to limited resource, the assumption that channel gains are known at the fusion center may be too strong. Thus, in this paper, we consider the case of noncoherent detection in the hybrid MAC WSN. Our goal here is to develop fusion rules which require only the knowledge of channel statistics instead of channel gains. Specifically, we present the optimal fusion rule based on the LR test and derive three suboptimal alternatives with low complexity, namely the weighed energy detector (WED), the deflection-coefficient-maximization (DCM), and the two-step (TS) rules.

The remainder of the paper is organized as follows. In Section 2, we formulate the fusion problem in the hybrid MAC WSN. In Section 3, we derive the optimal LR-based fusion rule and three sub-optimal alternatives. Performance analysis is contained in Section 4. Finally, some concluding remarks and future work are presented in Section 5.

Notation—Lower-case bold letters denote vectors; $E[\cdot]$, $Var[\cdot]$, $D[\cdot]$, and $(\cdot)^T$ are used to denote expectation, variance, deflection coefficient, and transpose, respectively; $P(\cdot)$, $p(\cdot)$ denote probability mass functions and probability density functions (pdf), in particular P(A|B) and p(a|b) represent the probability of event A conditioned on event B and the pdf of random variable a conditioned on random variable a, respectively.

2. System Model

Consider a sensor network with N sensors, where each sensor collects its observation generated according to either H_0 (e.g., the target is absent) or H_1 (e.g., the target is present), which are two hypotheses being tested. The prior probabilities of H_0 and H_1 are denoted by $P(H_0)$ and $P(H_1)$, respectively. After receiving its observation, each sensor makes a binary decision and transmits it with ON-OFF signaling to the FC. The model of the distributed detection system is illustrated in Figure 1. The N sensors are divided into K groups, with the kth group containing L_k sensors, and $N = \sum_{k=1}^K L_k$. Denote $S_{k,l}$, $l=1,\ldots,L_k$ as the lth sensor in the kth group, and $u_{k,l} \in \{0,1\}$ as the binary decision made by $S_{k,l}$. For sensor $S_{k,l}$, it will transmit a pulse (i.e., $u_{k,l} = 1$) to the FC if H_1 is decided, and will remain silent (i.e., $u_{k,l} = 0$) during its transmission period if H_0 is decided. The detection performance of sensor $S_{k,l}$ can be characterized by its detection probability:

$$P_{d_{k,l}} = P(u_{k,l} = 1|H_1) (1)$$

and its false alarm probability:

$$P_{f_{k,l}} = P(u_{k,l} = 1|H_0). (2)$$

In hybrid MAC networks, sensors in each group transmit their decisions via the same channel (i.e., the same time slot or frequency band) and different groups communicate with the FC through independent and mutually orthogonal channels. Thus, the received signal at the FC can be expressed as

$$y_k = \sum_{l=1}^{L_k} h_{k,l} e^{j\phi_{k,l}} u_{k,l} + n_k, \quad k = 1, \dots, K,$$
 (3)

where $h_{k,l}$ is the real-valued channel gain, $\phi_{k,l}$ is the channel phase, and n_k is zero-mean complex Gaussian noise with variance $2\sigma_k^2$.

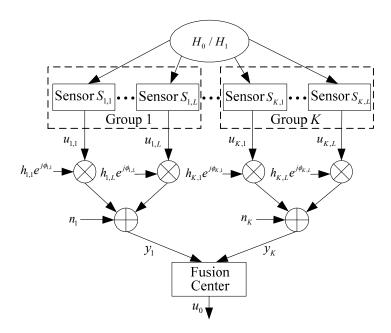


Figure 1. System model.

3. Noncoherent Decision Fusion

Noncoherent reception is a useful technique because it does not require the knowledge of channel phase. Using such technique, the FC can employ fusion rules based on the received signal envelop, or equivalently, the signal power. Therefore, we consider the case of noncoherent reception at the FC and develop fusion rules based on the received signal power $\left\{z_k = |y_k|^2, k = 1, \ldots, K\right\}$.

3.1. LR-Based Fusion Rule

The optimum fusion rule can be formulated based on the log-likelihood ratio test:

$$\Lambda_{LR} = \ln \frac{p(z_1, \dots, z_K | H_1)}{p(z_1, \dots, z_K | H_0)} \underset{H_0}{\overset{H_1}{\geqslant}} \gamma, \tag{4}$$

where the threshold γ can be determined from the false alarm constraint in the Neyman–Pearson test or can be chosen to minimize the fusion error probability in the Bayesian test [39].

Define a random variable $M_k \triangleq M(\mathbf{u}_k) = \sum_{l=1}^{L_k} u_{k,l}$, where $\mathbf{u}_k = (u_{k,1}, u_{k,2}, \cdots, u_{k,L_k})^T$. Assuming a Rayleigh fading channel with unit power, we can obtain [27]

$$p(z_k|M_k = m) = \frac{1}{m + 2\sigma_k^2} e^{-\frac{z_k}{m + 2\sigma_k^2}}.$$
 (5)

Subsequently, we have

$$p(z_k|H_i) = \sum_{m=0}^{L_k} p(z_k|M_k = m) P(M_k = m|H_i), \quad i = 0, 1,$$
(6)

Sensors **2019**, 19, 120 4 of 15

where $P(M_k = m|H_i)$ can be represented by the Poisson-binomial distribution, namely

$$P(M_k = m|H_0) = \sum_{u_k: M(u_k) = m} \prod_{l=1}^{L_k} (P_{f_{k,l}})^{u_{k,l}} \prod_{s=1}^{L_k} (1 - P_{f_{k,s}})^{1 - u_{k,s}}, \tag{7}$$

$$P(M_k = m|H_1) = \sum_{u_k: M(u_k) = m} \prod_{l=1}^{L_k} (P_{d_{k,l}})^{u_{k,l}} \prod_{s=1}^{L_k} (1 - P_{d_{k,s}})^{1 - u_{k,s}}.$$
 (8)

Further, if local decisions are independent and identically distributed (i.i.d.), i.e., $P_{f_{k,l}} = P_f$ and $P_{d_{k,l}} = P_d$, we have

$$P(M_k = m|H_0) = \binom{L_k}{m} \left(P_f\right)^m \left(1 - P_f\right)^{L_k - m},\tag{9}$$

$$P(M_k = m|H_1) = \binom{L_k}{m} (P_d)^m (1 - P_d)^{L_k - m}.$$
 (10)

Substituting (5), (9) and (10) into (6), and using (4), we can obtain the optimal LR-based fusion statistic:

$$\Lambda_{LR} = \ln \frac{p(z_1, z_2, \cdots, z_K | H_1)}{p(z_1, z_2, \cdots, z_K | H_0)} \\
= \sum_{k=1}^K \ln \frac{\sum_{m=0}^{L_k} \binom{L_k}{m} \frac{(P_d)^m (1 - P_d)^{L_k - m}}{m + 2\sigma_k^2} e^{-\frac{z_k}{m + 2\sigma_k^2}}}{\sum_{m=0}^{L_k} \binom{L_k}{m} \frac{(P_f)^m (1 - P_f)^{L_k - m}}{m + 2\sigma_k^2} e^{-\frac{z_k}{m + 2\sigma_k^2}}}$$
(11)

where the assumption of conditional independence of z_k , k = 1, ..., K, is used.

We can see that, as the number of groups increases, the test statistic in Equation (11) via sums of exponential functions becomes more and more complicated. Since most WSNs are resource constrained with regards to energy and processing capabilities, the optimal LR-based fusion rule with high complexity may be unsuitable for many practical WSN applications. Suboptimal alternatives with reduced complexity are then more desirable.

3.2. WED Fusion Rule

In this part, we consider the low-SNR approximation for the LR-based fusion rule and have the following proposition.

Proposition 1. As the channel noise variance $\sigma_k^2 \to \infty$, the LR-based fusion statistic in Equation (11) reduces to a form analogous to a WED statistic:

$$\Lambda_{WED} = \sum_{k=1}^{K} \sum_{m=0}^{L_k} {L_k \choose m} \frac{\left(P_f\right)^m \left(1 - P_f\right)^{L_k - m} - \left(P_d\right)^m \left(1 - P_d\right)^{L_k - m}}{m + 2\sigma_k^2} z_k.$$
 (12)

Sensors **2019**, 19, 120 5 of 15

Proof. For low SNR, i.e., $\sigma_k^2 \to \infty$,

$$\begin{split} \Lambda_{LR} &\approx \sum_{k=1}^{K} \ln \frac{\sum\limits_{m=0}^{L_k} \binom{L_k}{m} (P_d)^m (1-P_d)^{L_k-m} e^{-\frac{z_k}{m+2\sigma_k^2}}}{\sum\limits_{m=0}^{L_k} \binom{L_k}{m} (P_f)^m (1-P_f)^{L_k-m} e^{-\frac{z_k}{m+2\sigma_k^2}}} \\ &\stackrel{(a)}{\approx} \sum\limits_{k=1}^{K} \ln \frac{\sum\limits_{m=0}^{L_k} \binom{L_k}{m} (P_d)^m (1-P_d)^{L_k-m} (1-\frac{z_k}{m+2\sigma_k^2})}{\sum\limits_{m=0}^{L_k} \binom{L_k}{m} (P_f)^m (1-P_f)^{L_k-m} (1-\frac{z_k}{m+2\sigma_k^2})} \\ &= \sum\limits_{k=1}^{K} \ln \frac{1-\sum\limits_{m=0}^{L_k} \binom{L_k}{m} (P_f)^m (1-P_d)^{L_k-m} \frac{z_k}{m+2\sigma_k^2}}{1-\sum\limits_{m=0}^{L_k} \binom{L_k}{m} (P_f)^m (1-P_f)^{L_k-m} \frac{z_k}{m+2\sigma_k^2}} \\ &\stackrel{(b)}{\approx} \sum\limits_{k=1}^{K} \sum\limits_{m=0}^{L_k} \binom{L_k}{m} \left[\binom{P_f}{m} (1-P_f)^{L_k-m} - (P_d)^m (1-P_d)^{L_k-m} \right] \frac{z_k}{m+2\sigma_k^2} \end{split}$$

where we have used $e^x \approx 1 + x$ for small x in (a) and $\ln(1+x) \approx x$ for small x in (b). \square

Notice that the WED fusion statistic in Equation (12) is a linear combination of the received powers z_k/s . Under the condition that sensors are uniformly grouped and all the channels have the same noise power, i.e., $L_k = L$ and $\sigma_k^2 = \sigma^2$ for all k, the WED fusion statistic can reduce to a form analogous to a simple ED statistic:

$$\Lambda_{ED} = \sum_{k=1}^{K} z_k. \tag{13}$$

3.3. Decision Fusion Rule via Maximization of Deflection Coefficient

Different from Section 3.2, in this part, we will obtain the combining weights of the linear-combining fusion statistic by maximizing the deflection coefficient [11]. The deflection coefficient could reflect the output SNR and is often used to characterize the performance of a binary hypothesis test. It is worth noting that the use of deflection criterion usually yields a robust performance in many detection problems [12]. The deflection coefficient is defined as

$$D(\Lambda) = \frac{\left[E(\Lambda | H_1) - E(\Lambda | H_0) \right]^2}{\text{Var}(\Lambda | H_0)},\tag{14}$$

where $E(\cdot)$ and $Var(\cdot)$ denote the mean and variance of the fusion statistic, respectively. Using Equation (14), we have the following proposition.

Proposition 2. *Under the criterion of maximizing the deflection coefficient, the linear-combining fusion statistic, referred to as the DCM statistic, is given by*

$$\Lambda_{DCM} = \sum_{k=1}^{K} \frac{L_k z_k}{L_k^2 P_f^2 + L_k \left[(2 + 4\sigma_k^2) P_f - 2P_f^2 \right] + 4\sigma_k^4}.$$
 (15)

Proof. See the Appendix A. \Box

It is clear that when sensors are uniformly grouped and all the channels have the same noise power, the DCM fusion statistic is equivalent to the ED statistic. Thus, the DCM and WED fusion

Sensors **2019**, 19, 120 6 of 15

statistics coincide in this special case. It is worth noting that the DCM fusion statistic only requires the mean and the variance, which makes it very convenient to use for many practical applications.

3.4. Two-Step Fusion Rule

Motivated by the decode-then-fuse approach proposed in Equation [29], we now consider the case of high-channel SNR and develop the two-step fusion statistic as follows.

In the first step, each group makes a binary decision \widetilde{u}_k based on z_k . This is equivalent to a distributed detection using the nonorthogonal MAC, and the decision \widetilde{u}_k can be obtained through the maximum likelihood (ML) detector [39] as

$$p(z_k|H_1) \underset{\widetilde{u}_k=0}{\overset{\widetilde{u}_k=1}{\geqslant}} p(z_k|H_0). \tag{16}$$

As shown in Equation [28], this ML detector is equivalent to a simple threshold test on z_k . Therefore, we have

$$\widetilde{u}_k = I(z_k \ge \tau_k),\tag{17}$$

where $I(\cdot)$ is the indicator function, and the threshold τ_k of the kth group can be calculated by

$$\sum_{m=0}^{L_k} {L_k \choose m} \frac{\left(P_f\right)^m \left(1 - P_f\right)^{L_k - m} - \left(P_d\right)^m \left(1 - P_d\right)^{L_k - m}}{m + 2\sigma_k^2} e^{-\frac{\tau_k}{m + 2\sigma_k^2}} = 0.$$
 (18)

Define $\tilde{\mathbf{u}} = \{\tilde{u}_k, k = 1, ..., K\}$ and the probabilities of false alarm and detection of the kth group by P_{F_k} and P_{D_k} , respectively. Thus, we have

$$P(\widetilde{\mathbf{u}}|H_0) = \prod_{z_k > \tau_k} P(\widetilde{u}_k = 1|H_0) \prod_{z_k < \tau_k} P(\widetilde{u}_k = 0|H_0)$$

$$= \prod_{z_k > \tau_k} P_{F_k} \prod_{z_k < \tau_k} (1 - P_{F_k})$$
(19)

$$P(\widetilde{\mathbf{u}}|H_1) = \prod_{z_k > \tau_k} P(\widetilde{u}_k = 1|H_1) \prod_{z_k < \tau_k} P(\widetilde{u}_k = 0|H_1)$$

$$= \prod_{z_k > \tau_k} P_{D_k} \prod_{z_k < \tau_k} (1 - P_{D_k})$$
(20)

where P_{F_k} and P_{D_k} can be obtained by

$$P_{F_k} = \sum_{m=0}^{L_k} \binom{L_k}{m} (P_f)^m (1 - P_f)^{L_k - m} e^{-\frac{\tau_k}{m + 2\sigma_k^2}}, \tag{21}$$

$$P_{D_k} = \sum_{m=0}^{L_k} \binom{L_k}{m} (P_d)^m (1 - P_d)^{L_k - m} e^{-\frac{\tau_k}{m + 2\sigma_k^2}}.$$
 (22)

In the second step, the fusion center makes a global decision u_0 based on the likelihood ratio test on $\tilde{\mathbf{u}}$:

$$\Lambda_{TS} = \ln \frac{P(\widetilde{\mathbf{u}}|H_1)}{P(\widetilde{\mathbf{u}}|H_0)} = \sum_{z_k > \tau_k} \ln \frac{P_{D_k}}{P_{F_k}} + \sum_{z_k < \tau_k} \ln \frac{1 - P_{D_k}}{1 - P_{F_k}}.$$
 (23)

Note that for high-channel SNR, all the groups make correct detection with a large likelihood and this two-step fusion statistic, referred to as the TS statistic, should be a good approximation for the optimal LR-based fusion statistic.

Sensors **2019**, 19, 120 7 of 15

4. Performance Analysis

4.1. Asymptotic Behaviors of Linear-Combining Fusion Rules

In the following, we evaluate asymptotic behaviors for uniformly grouped hybrid MAC. We have shown in Section 3 that, in the case of the same channel SNR, the proposed linear-combining statistics (i.e., WED and DCM statistics) will reduce to a simple ED statistic. Thus, in this part, we focus on this ED statistic and evaluate its asymptotic behaviors as $K \to \infty$ and $L \to \infty$, respectively.

4.1.1. Asymptotic Behavior as $K \to \infty$

When $L_k = L$ and $\sigma_k^2 = \sigma^2$ for all k, the ED fusion statistic in Equation (13) is the sum of i.i.d. random variables. Therefore, if the number of groups (i.e., K) is large, this fusion statistic can be approximated by Gaussian distribution according to the central limit theorem. Thus, we have

$$\Lambda_{ED} \begin{cases}
\sim \left(\mu_0, \sigma_{ED_0}^2\right), & \text{under } H_0 \\
\sim \left(\mu_1, \sigma_{ED_1}^2\right), & \text{under } H_1
\end{cases} ,$$
(24)

where $X \sim (\mu, \sigma_{ED}^2)$ denotes that X is a Gaussian random variable with mean μ and variance σ_{ED}^2 , and where μ_i and $\sigma_{ED_i}^2$, for i = 0,1, are given by

$$\mu_0 = NP_f + 2K\sigma^2,\tag{25}$$

$$\sigma_{ED_0}^2 = (L-2)NP_f^2 + (2+4\sigma^2)NP_f + 4K\sigma^4, \tag{26}$$

$$\mu_1 = NP_d + 2K\sigma^2,\tag{27}$$

$$\sigma_{ED_1}^2 = (L-2)NP_d^2 + (2+4\sigma^2)NP_d + 4K\sigma^4.$$
(28)

4.1.2. Asymptotic Behavior as $L \to \infty$

When the group size (i.e., *L*) is large, we have the following proposition.

Proposition 3. As $L \to \infty$, the system-level probabilities of detection and false alarm are, respectively

$$P_{D_0} = e^{-\frac{\gamma}{LP_d + 2\sigma^2}} \sum_{i=0}^{K-1} \frac{1}{i!} \left(\frac{\gamma}{LP_d + 2\sigma^2}\right)^i$$
 (29)

and

$$P_{F_0} = e^{-\frac{\gamma}{LP_f + 2\sigma^2}} \sum_{i=0}^{K-1} \frac{1}{i!} \left(\frac{\gamma}{LP_f + 2\sigma^2} \right)^i, \tag{30}$$

where γ is a threshold.

Proof. As $L \to \infty$, the conditional probability density function of z_k can be approximated by Ref. [27]

$$p(z_k|H_1) = \frac{1}{LP_d + 2\sigma^2} e^{-\frac{-\zeta_k}{LP_d + 2\sigma^2}},$$
(31)

$$p(z_k|H_0) = \frac{1}{LP_f + 2\sigma^2} e^{-\frac{z_k}{LP_f + 2\sigma^2}}.$$
 (32)

Since Λ_{ED} in Equation (13) is the sum of random variables obeying the exponential distribution, we can obtain

$$p(\Lambda_{ED} = x | H_1) = \frac{x^{K-1}}{(K-1)!(LP_d + 2\sigma^2)^K} e^{-\frac{x}{LP_d + 2\sigma^2}},$$
(33)

$$p(\Lambda_{ED} = x | H_0) = \frac{x^{K-1}}{(K-1)!(LP_f + 2\sigma^2)^K} e^{-\frac{x}{LP_f + 2\sigma^2}}.$$
 (34)

With the above results, we can derive the system-level probability of detection:

$$P_{D_{0}} = \int_{\gamma}^{\infty} p(\Lambda_{ED} = x | H_{1}) dx$$

$$= \frac{1}{(K-1)! (LP_{d} + 2\sigma^{2})^{K}} e^{-\frac{\gamma}{LP_{d} + 2\sigma^{2}}} \sum_{i=0}^{K-1} i! {K-1 \choose i} \gamma^{K-1-i} (LP_{d} + 2\sigma^{2})^{i+1}$$

$$= e^{-\frac{\gamma}{LP_{d} + 2\sigma^{2}}} \sum_{i=0}^{K-1} \frac{1}{i!} (\frac{\gamma}{LP_{d} + 2\sigma^{2}})^{i}$$
(35)

In the above derivation, we have used the formula in Ref. ([40], Equation (2.321)). Similarly, we have the system-level false alarm rate as in Equation (30). \Box

4.2. Performance Analysis of the TS Fusion Rule

In evaluating the performance of the TS fusion rule, it is assumed that sensors are uniformly grouped and all the channels have the same noise power. The system-level probabilities of false alarm, detection, and error are provided in the following proposition.

Proposition 4.

(a) For the Neyman-Pearson detection, the system-level probabilities of false alarm and detection are given by

$$P_{F_0} = \sum_{i=K_{\tau}}^{K} \binom{K}{i} P_F^i (1 - P_F)^{K-i}, \tag{36}$$

$$P_{D_0} = \sum_{i=K_{\tau}}^{K} \binom{K}{i} P_D^i (1 - P_D)^{K-i}, \tag{37}$$

where K_{τ} is a discrete threshold.

(b) For the Bayesian detection, the fusion error probability is given by

$$P_e = P(H_1) - P(H_1) \sum_{i=T}^{K} {K \choose i} P_D^i (1 - P_D)^{K-i} + P(H_0) \sum_{i=T}^{K} {K \choose i} P_F^i (1 - P_F)^{K-i}.$$
(38)

The threshold T can be obtained as $T = \lceil T^* \rceil$ *, where*

$$T^* = \frac{\ln\left\{\frac{P(H_0)}{P(H_1)} \left(\frac{1 - P_F}{1 - P_D}\right)^K\right\}}{\ln\{P_D(1 - P_F)/P_F(1 - P_D)\}}$$
(39)

and $\lceil \cdot \rceil$ denotes the standard ceiling function.

Proof. Under the condition that $L_k = L$ and $\sigma_k^2 = \sigma^2$ for all k, we have $P_{D_k} = P_D$, $P_{F_k} = P_F$, and $\tau_k = \tau$. Define $K_1 = |S_1|$, where $S_1 = \{k : z_k > \tau\}$, i.e., K_1 is the cardinality of S_1 . Thus, the TS fusion rule reduces to a simple counting rule:

$$\Lambda_{TS} = K_1 \ln \frac{P_D(1 - P_F)}{P_F(1 - P_D)} + K \ln \frac{(1 - P_D)}{(1 - P_F)}.$$
 (40)

Sensors **2019**, 19, 120 9 of 15

Note that the TS statistic is equivalent to K_1 , which follows binomial distribution, i.e., $K_1|H_1 \sim B(K, P_D)$ and $K_1|H_0 \sim B(K, P_F)$. Given K_τ , we obtain (36) and (37).

If the prior probabilities of H_0 and H_1 are known at the fusion center, we can employ the Bayesian approach to minimize the fusion error probability. Thus, we have

$$\Lambda_{TS} \underset{H_0}{\gtrless} \frac{P(H_0)}{P(H_1)}. \tag{41}$$

The fusion error probability can therefore be obtained as

$$P_e = P(H_0)P_{F_0} + P(H_1)[1 - P_{D_0}]. (42)$$

Substituting P_{F_0} and P_{D_0} into (42), we then obtain (38). Note that (39) can be derived by starting from (41) and exploiting the fact that the TS fusion statistic in Equation (40) is monotonic.

4.3. Numerical Results

In this part, we show some simulation results. We consider the simplified scenario where sensors collaborate to detect a known parameter in Gaussian noise with zero mean and unitary variance. Each sensor makes its binary decision by employing a threshold test on the observation. Its false alarm rate is set to $P_f = 0.05$ while the detection probability is $P_d = 0.5$. The channel SNR is defined as $\text{SNR}_k \triangleq 10 \log_{10}(1/(2\sigma_k^2))$ dB. All the simulation results are obtained by averaging over 10^5 realizations.

First, we consider the case where 30 sensors are partitioned into 10 groups, and different groups have different channel SNRs for their channels to the FC. Specifically, the sensor grouping strategy is denoted by $\{L_k\} = \{1, 1, 1, 3, 3, 3, 3, 5, 5, 5\}$, respectively. We assume that the channel SNRs are distributed as $\{SNR_k\} = \{S-6, S-4, S-2, S, S, S, S, S+2, S+4, S+6\}$ dB, respectively, where S is the arithmetic mean of all the channel SNRs. Figure 2 presents the receiver operating characteristic (ROC) curves for different fusion rules at S=0 dB. Here, an idealistic centralized detection scenario, where sensors transmit their raw observations to the FC, serves as a performance baseline for distributed detection. As shown, the LR-based fusion rule performs closely to the centralized detection, and provides the optimal performance among all the distributed fusion rules. The DCM fusion rule is slightly worse than the optimal LR-based fusion rule and outperforms both WED and TS fusion rules.

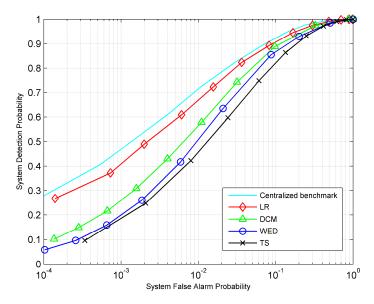


Figure 2. Receiver operating characteristic (ROC) curves for different fusion rules at S = 0 dB.

To better understand the performance difference as a function of channel SNR, Figure 3 gives the probability of detection as a function of the mean value of channel SNRs for various fusion rules. The system-level false alarm rate is fixed at $P_{F_0} = 0.01$ (unless otherwise specified). From this figure, it is clear that the performances of WED and TS fusion rules approach that of the LR-based fusion rule at very low and very high SNR respectively. The simulation results confirm that the WED fusion rule is a low-SNR approximation of the LR-based fusion rule, and the TS fusion rule is a high-SNR approximation of the LR-based rule. However, when the channel SNR is medium, the DCM fusion rule has the best performance among the three sub-optimal rules. Table 1 gives the performance rating for the proposed fusion rules under different SNR range. In addition, we report the comparison of detection performance, computational complexity, and the requirement for a priori information for each of the fusion rules in Table 2.

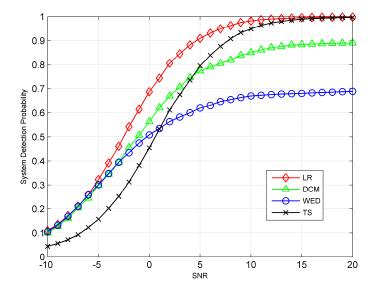


Figure 3. The probability of detection as a function of the mean value of channel SNRs for various fusion rules.

Table 1. Performance rating under different SNR range.

Mean Value of Channel SNRs (dB)	Performance Rating
-10~-3	LR > WED > DCM > TS
-2~1	LR > DCM > WED > TS
2~4	LR > DCM > TS > WED
5~20	LR > TS > DCM > WED

Table 2. Comparison of the fusion rules.

Fusion Rule	Required Parameters	Complexity	Performance
LR	P_d , P_f , σ_k^2	Most complex	Optimum
WED	P_d , P_f , σ_k^2	Simple	Near-optimal for low SNR
DCM	P_f, σ_k^2	Most simple	Near-optimal for moderate SNR
TS	P_d , P_f , σ_k^2	Complex	Near-optimal for high SNR

Next, we analyze the detection performance of the hybrid MAC under a uniform sensor grouping strategy. We assume that all the channels have the same SNR. Figure 4 gives the probability of detection as a function of channel SNR for the ED fusion rule. Figure 5 gives the numerically computed (i.e., using Equation (38)) error probability as a function of channel SNR for the TS fusion rule. Specifically, the prior probabilities of H_0 and H_1 are assumed to be equally likely. As shown in Figures 4 and 5, the performance of the hybrid MAC is bounded by those of orthogonal and nonorthogonal MACs

when the channel SNR is high or sufficiently low. In this case, the detection performances of both rules improve with the increase of group size when the channel SNR is low, and deteriorate with the increase of group size when the channel SNR is high. Note that, compared with the conventional orthogonal and non-orthogonal MACs, the hybrid MAC can provide a better ROC when the channel SNR is medium.

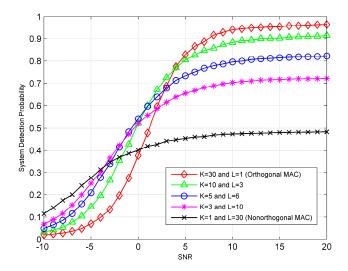


Figure 4. The probability of detection as a function of channel SNR for the ED fusion rule.

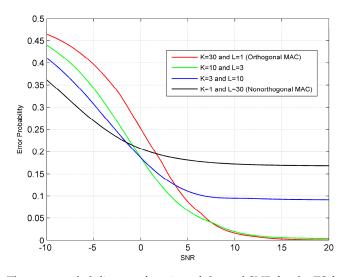


Figure 5. The error probability as a function of channel SNR for the TS fusion rule.

Finally, we examine the quality of approximation for the ED fusion rule. Specifically, we set SNR = -5 dB. Figure 6 shows both the Gaussian approximation and the simulation results for a large number of groups. When the number of groups is 100, the Gaussian approximation is seen to provide a tight fit to the actual ROC for all values of L tested. We can also observe that increasing L results in a decrease in the gap between the approximation and simulation results. Figure 7 gives the ROC curves obtained by using the approximations (29) and (30), and those by simulations for L = 100. As shown, the approximation results agree very well with the simulation results. In particular, when K = 3, we see the approximation and simulation curves merging as shown in Figure 7.

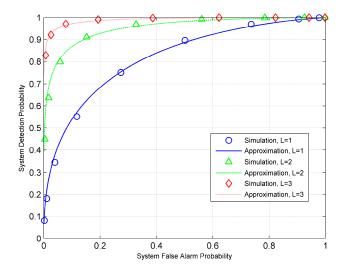


Figure 6. ROC curves for the ED fusion rule with K = 100.

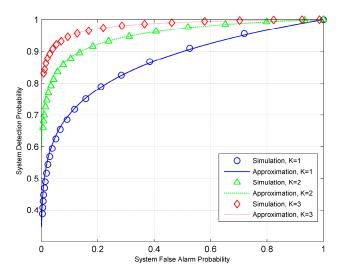


Figure 7. ROC curves for the ED fusion rule with L = 100.

5. Conclusions

In this work, we study the problem of decision fusion under the hybrid MAC scheme. Considering noncoherent detection at the fusion center, we present the optimal decision fusion rule based on the LR test and derive three sub-optimal rules with low-complexity. When sensors are uniformly grouped and all the channels have the same noise power, the WED and DCM fusion rules reduce to the ED fusion rule. For the ED rule, we derive closed-form results to the distribution of its fusion statistic under a large K and under a large L, respectively. In addition, we derive explicit formulas of the TS fusion statistic under the Neyman–Pearson criterion and the minimum probability of error criterion, respectively. Simulation results show that the WED fusion rule is a low-SNR approximation of the LR-based fusion rule, and the TS fusion rule is a high-SNR approximation of the LR-based fusion rule. The detection performance of the hybrid MAC generally improves with the increase of group size when the channel SNR is low, and deteriorates with the increase of group size when the channel SNR is high. Compared with the conventional orthogonal and nonorthogonal MACs, the hybrid MAC can achieve a better performance when the channel SNR is medium. In this work, we address only the fusion rule design for distributed detection of a known parameter in Gaussian noise. The detection of an unknown deterministic signal or a random time-correlated signal in noise is worth further investigation.

Author Contributions: Conceptualization, W.C. and S.L.; Methodology, K.W. and S.L.; Software, S.L.; Validation, K.W., and K.L.; Formal analysis, S.L.; Writing—original draft preparation, S.L.; Writing—review and editing, K.W.; Funding acquisition, K.L.

Funding: This research was funded by the National Natural Science Foundation of China (Grant Nos. 51279151, 61701356 and 61672395) and the Major Project for the Technology Innovation of Hubei Province, China, under Grant No. 2017AAA120.

Conflicts of Interest: The authors declare no conflict of interest.

Appendix A

Proof of Proposition 2. Let Λ be a linear combination of z_k 's. We have

$$\Lambda = \sum_{k=1}^{K} c_k z_k,\tag{A1}$$

where c_k 's denote the combining weights.

Using (6) and algebraic manipulation, we have

$$E[\Lambda|H_0] = \sum_{k=1}^{K} c_k (L_k P_f + 2\sigma_k^2),$$
(A2)

$$E[\Lambda|H_1] = \sum_{k=1}^{K} c_k (L_k P_d + 2\sigma_k^2),$$
(A3)

$$\operatorname{Var}[\Lambda|H_0] = \sum_{k=1}^{K} c_k^2 \left\{ L_k^2 P_f^2 + L_k \left[(2 + 4\sigma_k^2) P_f - 2P_f^2 \right] + 4\sigma_k^4 \right\}. \tag{A4}$$

It immediately follows from (A2)-(A4) that

$$D(\Lambda) = \frac{\left[\sum_{k=1}^{K} c_k L_k (P_d - P_f)\right]^2}{\sum_{k=1}^{K} c_k^2 \left\{ L_k^2 P_f^2 + L_k \left[(2 + 4\sigma_k^2) P_f - 2P_f^2 \right] + 4\sigma_k^4 \right\}}.$$
 (A5)

Denote

$$a_k \triangleq c_k \sqrt{L_k^2 P_f^2 + L_k \left[(2 + 4\sigma_k^2) P_f - 2P_f^2 \right] + 4\sigma_k^4}$$
 (A6)

and

$$b_k \triangleq \frac{L_k(P_d - P_f)}{\sqrt{L_k^2 P_f^2 + L_k \left[(2 + 4\sigma_k^2) P_f - 2P_f^2 \right] + 4\sigma_k^4}}.$$
 (A7)

Applying the Cauchy–Schwartz inequality, we have $D(\Lambda) \leq \sum_{k=1}^{K} b_k^2$, with the equality holding if and only if $a_k = \beta b_k$, where β is a constant.

Thus, the combining weights in (A1) that maximize $D(\Lambda)$ are given by

$$c_k = \beta \cdot \frac{L_k(P_d - P_f)}{L_k^2 P_f^2 + L_k \left[(2 + 4\sigma_k^2) P_f - 2P_f^2 \right] + 4\sigma_k^4}.$$
 (A8)

Substituting (A8) into (A1) and neglecting the constant term that does not affect detection performance, we obtain the DCM fusion statistic as in (15). \Box

References

1. Nevat, I.; Peters, G.W.; Collings, I.B. Distributed detection in sensor networks over fading channels with multiple antennas at the Fusion Centre. *IEEE Trans. Signal Process.* **2014**, *62*, 671–683. [CrossRef]

- 2. Khelifi, F.; Bradai, A.; Kaddachi, M.L.; Rawat, P. Design and experimental implementation of monitoring system in wireless sensor networks. *IET Wirel. Sens. Syst.* **2018**, *8*, 350–359. [CrossRef]
- 3. Ayaz, M.; Ammad-uddin, M.; Baig, I.; Aggoune, E.M. Wireless sensor's civil applications, prototypes, and future integration possibilities: A review. *IEEE Sens. J.* **2018**, *18*, 4–30. [CrossRef]
- 4. Lazarescu, M.T. Design and field test of a WSN platform prototype for long-term environmental monitoring. *Sensors* **2015**, *15*, 9481–9518. [CrossRef]
- 5. Popescu, D.; Dragana, C.; Stoican, F.; Ichim, L.; Stamatescu, G. A collaborative UAV-WSN network for monitoring large areas. *Sensors* **2018**, *18*, 4202. [CrossRef] [PubMed]
- 6. Yu, J.; Gong, W.; Liu, J.; Chen, L.; Wang, K.; Zhang, R. Missing tag identification in COTS RFID systems: Bridging the gap between theory and practice. *IEEE Trans. Mob. Netw.* **2018**. [CrossRef]
- 7. Yu, J.; Gong, W.; Liu, J.; Chen, L.; Wang, K. On efficient tree-based tag search in large-scale RFID systems. *IEEE/ACM Trans. Netw.* **2018**. [CrossRef]
- 8. Yan, Y.; Wang, H.; Shen, X.; Zhong, X. Decision fusion with channel errors in distributed decode-then-fuse sensor networks. *Sensors* **2015**, *15*, 19157–19180. [CrossRef] [PubMed]
- 9. Trasvina-Moreno, C.A.; Blasco, R.; Marco, A.; Casas, R.; Trasvina-Castro, A. Unmanned aerial vehicle based wireless sensor network for marine-coastal environment monitoring. *Sensors* **2017**, *17*, 460. [CrossRef]
- 10. Chair, Z.; Varshney, P.K. Optimal data fusion in multiple sensor detection systems. *IEEE Trans. Aerosp. Electron. Syst.* **1986**, 22, 98–101. [CrossRef]
- 11. Chen, B.; Jiang, R.; Kasetkasem, T.; Varshney, P.K. Channel aware decision fusion in wireless sensor networks. *IEEE Trans. Signal Process.* **2004**, *52*, 3454–3458. [CrossRef]
- 12. Niu, R.; Chen, B.; Varshney, P.K. Fusion of decisions transmitted over Rayleigh fading channels in wireless sensor networks. *IEEE Trans. Signal Process.* **2006**, *54*, 1018–1027.
- 13. Liu, K.; Yang, Y.; Jia, J. Fusion of decisions transmitted over flat fading channels via maximizing the deflection coefficient. *IEEE Trans. Veh. Technol.* **2010**, *59*, 3634–3640.
- 14. Cheng, V.; Wang, T. Performance analysis of distributed decision fusion using a censoring scheme in wireless sensor networks. *IEEE Trans. Veh. Technol.* **2010**, *59*, 2845–2851. [CrossRef]
- 15. Wang, T.; Wu, J. Does more transmitting sensors always mean better decision fusion in censoring sensor networks with an unknown size? *IEEE Trans. Commun.* **2012**, *60*, 2313–2324. [CrossRef]
- 16. Cheng, V.; Wang, T. Performance analysis of distributed decision fusion using a multilevel censoring scheme in wireless sensor networks. *IEEE Trans. Veh. Technol.* **2012**, *61*, 1610–1619. [CrossRef]
- 17. Cattivelli, F.S.; Sayed, A.H. Diffusion LMS strategies for distributed estimation. *IEEE Trans. Signal Process.* **2010**, *58*, 1035–1048. [CrossRef]
- 18. Chen, H.; Gao, F.; Martins, M.; Huang, P.; Liang, J. Accurate and efficient node localization for mobile sensor networks. *Mob. Netw. Appl.* **2013**, *18*, 141–147. [CrossRef]
- 19. Chen, H.; Liu, B.; Huang, P.; Liang, J.; Gu, Y. Mobility-assisted node localization based on TOA measurements without time synchronization in wireless sensor networks. *Mob. Netw. Appl.* **2012**, *17*, 90–99. [CrossRef]
- 20. Chen, H.; Sezaki, K. Distributed target tracking algorithm for wireless sensor networks. In Proceedings of the IEEE Conference on Communications (ICC), Kyoto, Japan, 5–9 June 2011; pp. 1–5.
- 21. Ciuonzo, D.; Salvo Rossi, P.; Willett, P. Generalized Rao test for decentralized detection of an uncooperative target. *IEEE Signal Process. Lett.* **2017**, 24, 678–682. [CrossRef]
- 22. Salvo Rossi, P.; Ciuonzo, D.; Ekman, T. HMM-based decision fusion in wireless sensor networks with noncoherent multiple access. *IEEE Commun. Lett.* **2015**, *19*, 871–874. [CrossRef]
- 23. Mergen, G.; Naware, V.; Tong, L. Asymptotic detection performance of type-based multiple access over multiaccess fading channels. *IEEE Trans. Signal Process.* **2007**, *55*, 1081–1092. [CrossRef]
- 24. Anandkumar, A.; Tong, L. Type-based random access for distributed detection over multiaccess fading channels. *IEEE Trans. Signal Process.* **2007**, *55*, 5032–5043. [CrossRef]
- 25. Liu, K.; Sayeed, A.M. Type-based decentralized detection in wireless sensor networks. *IEEE Trans. Signal Process.* **2007**, *55*, 1899–1910. [CrossRef]

26. Li, W.; Dai, H. Distributed detection in wireless sensor networks using a multiple access channel. *IEEE Trans. Signal Process.* **2007**, *55*, 822–833. [CrossRef]

- 27. Li, F.; Evans, J.S. Decision fusion over noncoherent fading multiaccess channels. In Proceedings of the 2008 IEEE Global Telecommunications Conference (GLOBECOM 2008), New Orleans, LA, USA, 1–4 December 2008; pp. 1–5.
- 28. Li, F.; Evans, J.S.; Dey, S. Decision fusion over noncoherent fading multiaccess channels. *IEEE Trans. Signal Process.* **2011**, *59*, 4367–4380. [CrossRef]
- 29. Ciuonzo, D.; Romano, G.; Salvo Rossi, P. Channel-aware decision fusion in distributed MIMO wireless sensor networks: Decode-and-fuse vs. decode-then-fuse. *IEEE Trans. Wirel. Commun.* **2012**, *11*, 976–2985. [CrossRef]
- 30. Ciuonzo, D.; Romano, G.; Salvo Rossi, P. Performance analysis of maximum ratio combining in channel-aware MIMO decision fusion. *IEEE Trans. Wirel. Commun.* **2013**, 12, 4716–4728. [CrossRef]
- 31. Ciuonzo, D.; Romano, G.; Salvo Rossi, P. Optimality of received energy in decision fusion over Rayleigh fading diversity MAC with non-identical sensors. *IEEE Trans. Signal Process.* **2013**, *61*, 22–27. [CrossRef]
- 32. Ciuonzo, D.; Aubry, A.; Carotenuto, V. Rician MIMO channel and jamming-aware decision fusion. *IEEE Trans. Signal Process.* **2017**, *65*, 3866–3880. [CrossRef]
- 33. Salvo Rossi, P.; Ciuonzo, D.; Kansanen, K.; Ekman, T. Performance analysis of energy detection for MIMO decision fusion in wireless sensor networks over arbitrary fading channels. *IEEE Trans. Wirel. Commun.* **2016**, 15, 7794–7806. [CrossRef]
- 34. Chen, Z.; Lin, T.; Wu, C. Decentralized learning-based relay assignment for cooperative communications. *IEEE Trans. Veh. Technol.* **2016**, *65*, 813–826. [CrossRef]
- 35. Al-Jarrah, M.; Al-Jarrah, R.; Al-Ababneh, N. Decision fusion in mobile wireless sensor networks using cooperative multiple symbol differential space time coding. *Int. J. Electron. Commun.* **2017**, *80*, 127–136. [CrossRef]
- 36. Al-Jarrah, M.; Al-Dweik, A.; Kalil, M.; Ikki, S. Efficient decision fusion for cooperative wireless sensor networks. In Proceedings of the 2017 International Conference on Electrical and Computing Technologies and Applications (ICECTA), Ras Al Khaimah, United Arab Emirates, 21–23 November 2017; pp. 1–5.
- 37. Liu, J.; Chung, C. Distributed estimation in a wireless sensor network using hybrid MAC. *IEEE Trans. Veh. Technol.* **2011**, *60*, 3424–3435. [CrossRef]
- 38. Liu, J.; Chou, P.; Chung, C. Distributed detection in a wireless sensor network using hybrid MAC. In Proceedings of the 2013 IEEE 77th Vehicular Technology Conference (VTC Spring), Dresden, Germany, 2–5 June 2013; pp. 1–5.
- 39. Kay, S.M. Fundamentals of Statistical Signal Processing: Detection Theory; Prentice-Hall: Upper Saddle River, NJ, USA, 1993.
- 40. Gradshteyn, I.S.; Ryzhik, I.M. *Table of Integrals, Series and Products*, 7th ed.; Academic: New York, NY, USA, 2007.



© 2019 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (http://creativecommons.org/licenses/by/4.0/).